

وزارة التعليم العالي والبحث العلمي جامعة المستقرل

# LEC : FIVE

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Lecture Title : Conversion Delta To Star Connection And

**Conversion Star To Delta Connection** 

Analysis Methods and Y-D (Star-Delta)

# **8.2 CURRENT SOURCES**

A current source determines the direction and magnitude of the current in the branch where it is located.

The magnitude and the polarity of the voltage across a current source are each a function of the network to which the voltage is applied.

**EXAMPLE 8.1** Find the source voltage, the voltage *V*<sub>1</sub>, and current *I*<sub>1</sub> for the circuit in Fig.

Since the current source establishes the current in the branch in which it is located, the current  $I_1$  must equal I, and

$$I_1 = I = 10 \text{ mA}$$

The voltage across  $R_1$  is then determined by Ohm's law:

$$V_1 = I_1 R_1 = (10 \text{ mA})(20 \text{ k}\Omega) = 200 \text{ V}$$

Since resistor  $R_1$  and the current source are in parallel,

$$V_s = V_1 = 200 \,\mathrm{V}$$



**EXAMPLE 8.2** Find the voltage *Vs and currents I*<sup>1</sup> *and I*<sup>2</sup> *for the network* in Fig.

for the current source, *Vs must be determined, and for the voltage source, Is must be* determined.



Since the current source and voltage source are in parallel,

$$V_s = E = 12 \text{ V}$$

Further, since the voltage source and resistor R are in parallel,

$$V_R = E = 12 \text{ V}$$

and

$$I_2 = \frac{V_R}{R} = \frac{12 \text{ V}}{4 \Omega} = 3 \text{ A}$$

The current  $I_1$  of the voltage source can then be determined by applying Kirchhoff's current law at the top of the network as follows:

$$\Sigma I_i = \Sigma I_o$$

$$I = I_1 + I_2$$

$$I_1 = I - I_2 = 7 \text{ A} - 3 \text{ A} = 4 \text{ A}$$

and

#### **8.3 SOURCE CONVERSIONS**

For the voltage source, if *Rs=0*, or if it is so small compared to any series resistors that it can be ignored, then we have an "ideal" voltage source for all practical purposes.

For the current source, since the resistor Rp is in parallel, if  $R_P = \infty$ , or if it is large enough compared to any parallel resistive elements that it can be ignored, then we have an "ideal" current source.

# $R_{s} \neq I_{L}$ $R_{L} \neq I_{L}$ $R_{L} \neq R_{L} \neq R_{L$

(a)

(b)



Source conversion.

## Note that:

1Ideal sources *cannot be converted from one* type to another.

2The equivalence between a current source and a voltage source exists only at their external terminals. **EXAMPLE 8.4** For the circuit in Fig.

a. Determine the current IL.

b. Convert the voltage source to a current source.c.Using the resulting current source of part (b), calculate the current through the load resistor, and compare your answer to the result of part (a).

a. Applying Ohm's law:

$$I_L = \frac{E}{R_s + R_L} = \frac{6 \text{ V}}{2 \Omega + 4 \Omega} = \frac{6 \text{ V}}{6 \Omega} = 1 \text{ A}$$

b. Using Ohm's law again:

$$I = \frac{E}{R_s} = \frac{6 \text{ V}}{2 \Omega} = 3 \text{ A}$$

c. Using the current divider rule:

$$I_L = \frac{R_p I}{R_p + R_L} = \frac{(2 \ \Omega)(3 \ A)}{2 \ \Omega + 4 \ \Omega}$$
$$= \frac{1}{3} (3 \ A) = 1 \ A$$





### **8.4 CURRENT SOURCES IN PARALLEL**

*Current sources of different values cannot be placed in series due to a violation of Kirchhoff's current law.* 

However, current sources can be placed in parallel just as voltage sources can be placed in series. In general,

Two or more current sources in parallel can be replaced by a single current source having a magnitude determined by the difference of the sum of the currents in one direction and the sum in the opposite direction. The new parallel internal resistance is the total resistance of the resulting parallel resistive elements.

# Current sources of different current ratings cannot be connected in series.

**EXAMPLE 8.6** Reduce the parallel current sources in Fig. to a single current source.



**EXAMPLE 8.8** Reduce the network in Fig. to a single current source, and calculate the current through R<sub>L</sub>.



$$I_s = I_1 + I_2 = 4 \text{ A} + 6 \text{ A} = \mathbf{10} \text{ A}$$
$$R_s = R_1 || R_2 = 8 \Omega || 24 \Omega = \mathbf{6} \Omega$$

Applying the current divider rule to the resulting network in Fig.

$$I_L = \frac{R_p I_s}{R_p + R_L} = \frac{(6 \ \Omega)(10 \ \text{A})}{6 \ \Omega + 14 \ \Omega} = \frac{60 \ \text{A}}{20} = 3 \ \text{A}$$



#### **8.6 BRANCH-CURRENT ANALYSIS**

**1**. Assign a distinct current of arbitrary direction to each branch of the network.

2. Indicate the polarities for each resistor as determined by the assumed current direction.

3. Apply Kirchhoff's voltage law around each closed, independent loop of the network.

**4.**Apply Kirchhoff's current law at the minimum number of nodes that will include all the branch currents of the network.

5. Solve the resulting simultaneous linear equations for assumed branch currents.

![](_page_7_Figure_6.jpeg)

Determining the number of independent closed loops.

![](_page_7_Figure_8.jpeg)

**EXAMPLE 8.9** Apply the branch-current method to the network in Fig.

Step 1: Since there are three distinct branches (cda, cba, ca), three currents of arbitrary directions (I1, I2, I3) are chosen, as shown. The current directions for 11 and 12 were chosen to match the "pressure" applied by sources E1 and E<sub>2</sub>, respectively.

Since both I1 and I2 enter node a, I3 is leaving.

![](_page_8_Figure_3.jpeg)

![](_page_8_Figure_4.jpeg)

by  $I_1$ 

*Step 3: Kirchhoff's voltage law is applied around each closed loop (1 and 2)* in the clockwise direction:

loop 1: 
$$\Sigma_{\mathbb{C}} V = +E_1 - V_{R_1} - V_{R_3} = 0$$
  
 $\uparrow$  Drop in potential  
loop 2:  $\Sigma_{\mathbb{C}} V = +V_{R_3} + V_{R_2} - E_2 = 0$   
and  $\uparrow$  Drop in potential  
loop 1:  $\Sigma_{\mathbb{C}} V = +2 V - (2 \Omega)I_1 - (4 \Omega)I_3 = 0$   
Battery Voltage drop Voltage drop  
potential across 2 \Omega across 4 \Omega  
resistor  
loop 2:  $\Sigma_{\mathbb{C}} V = (4 \Omega)I_3 + (1 \Omega)I_2 - 6 V = 0$ 

Step 4: Applying Kirchhoff's current law at node a (in a two-node network, the law is applied at only one node),

$$I_1 + I_2 = I_3$$

Step 5: There are three equations and three unknowns (units removed for clarity):

Using third-order determinants (Appendix D), we have

![](_page_10_Figure_3.jpeg)

**Solution 2:** Instead of using third-order determinants as in Solution 1, we can reduce the three equations to two by substituting the third equation in the first and second equations:

$$\underbrace{\begin{array}{c}I_{3}\\2-2I_{1}-4(I_{1}+I_{2})=0\\I_{3}\\4(I_{1}+I_{2})+I_{2}-6=0\end{array}}_{I_{3}}\left\{\begin{array}{c}2-2I_{1}-4I_{1}-4I_{2}=0\\4I_{1}+4I_{2}+I_{2}-6=0\\-6I_{1}-4I_{2}=-2\\+4I_{1}+5I_{2}=+6\end{array}\right\}$$

Multiplying through by -1 in the top equation yields

$$6I_1 + 4I_2 = +24I_1 + 5I_2 = +6$$

and using determinants,

or

$$I_1 = \frac{\begin{vmatrix} 2 & 4 \\ 6 & 5 \end{vmatrix}}{\begin{vmatrix} 6 & 4 \\ 4 & 5 \end{vmatrix}} = \frac{10 - 24}{30 - 16} = \frac{-14}{14} = -1 \text{ A}$$

#### Mesh Analysis Procedure

- 1. Assign a distinct current in the clockwise direction to each independent, closed loop of the network. It is not absolutely necessary to choose the clockwise direction for each loop current. However, by choosing the clockwise direction as a standard, we can develop a shorthand method (Section 8.8) for writing the required equations that will save time and possibly prevent some common errors.
- 2. Indicate the polarities within each loop for each resistor as determined by the assumed direction of loop current for that loop. Note the requirement that the polarities be placed within each loop. This requires, as shown in Fig., that the 4  $\Omega$  resistor have two sets of polarities across it.
- **3.** Apply Kirchhoff's voltage law around each closed loop in the clockwise direction. Again, the clockwise direction was chosen to establish uniformity and prepare us for the method to be introduced in the next section.
  - a. If a resistor has two or more assumed currents through it, the total current through the resistor is the assumed current of the loop in which Kirchhoff's voltage law is being applied, plus the assumed currents of the other loops passing through in the same direction, minus the assumed currents through in the opposite direction.
  - b. The polarity of a voltage source is unaffected by the direction of the assigned loop currents.
- 4. Solve the resulting simultaneous linear equations for the assumed loop currents.

![](_page_13_Figure_0.jpeg)

loop 2: 
$$-V_3 - V_2 - E_2 = 0$$
 (clockwise starting at b)  
 $-(4 \Omega)(I_2 - I_1) - (1 \Omega)I_2 - 6 V = 0$ 

The equations are then rewritten as follows

loop 1: 
$$+2 - 2I_1 - 4I_1 + 4I_2 = 0$$
  
loop 2:  $-4I_2 + 4I_1 - 1I_2 - 6 = 0$ 

Applying determinants results in

$$I_1 = -1 A$$
 and  $I_2 = -2 A$ 

The actual current through the 2 V source and 2  $\Omega$  resistor is therefore 1 A in the other direction, and the current through the 6 V source and 1  $\Omega$  resistor is 2 A in the opposite direction indicated on the circuit.

The current through the 4  $\Omega$  resistor is determined by the following equation from the original network:

loop 1: 
$$I_{4\Omega} = I_1 - I_2 = -1 \text{ A} - (-2 \text{ A}) = -1 \text{ A} + 2 \text{ A}$$
  
= 1 A (in the direction of  $I_1$ )

# **8.9 NODAL ANALYSIS (GENERAL APPROACH)**

The number of nodes for which the voltage must be determined using nodal analysis is 1 less than the total number of nodes.

**Nodal Analysis Procedure** 

**1.** Determine the number of nodes within the network.

2.Pick a reference node, and label each remaining node with a subscripted value of voltage: V1, V2, and so on.

**3.**Apply Kirchhoff's current law at each node except the reference.

Assume that all unknown currents leave the node for each application of Kirchhoff's current law. In other words, for each node, don't be influenced by the direction that an unknown current for another node may have had. Each node is to be treated as a separate entity, independent of the application of Kirchhoff's current law to the other nodes.

4. Solve the resulting equations for the nodal voltages.

**EXAMPLE 8.21** Determine the nodal voltages for the network shown.

For the node V1. Applying Kirchhoff's current law:

and 
$$A = I_1 + I_3$$
  
and  $A = \frac{V_1}{R_1} + \frac{V_1 - V_2}{R_3} = \frac{V_1}{2\Omega} + \frac{V_1 - V_2}{12\Omega}$ 

Expanding and rearranging:

$$V_1\left(\frac{1}{2\Omega} + \frac{1}{12\Omega}\right) - V_2\left(\frac{1}{12\Omega}\right) = 4 \text{ A}$$

For node V2, the currents are defined as

$$0 = I_3 + I_2 + 2 A$$
  
and 
$$\frac{V_2 - V_1}{R_3} + \frac{V_2}{R_2} + 2 A = 0$$
$$\frac{V_2 - V_1}{12 \Omega} + \frac{V_2}{6 \Omega} + 2 A = 0$$

Expanding and rearranging:

$$V_2\left(\frac{1}{12\,\Omega} + \frac{1}{6\,\Omega}\right) - V_1\left(\frac{1}{12\,\Omega}\right) = -2\,\mathrm{A}$$

![](_page_16_Figure_9.jpeg)

resulting in Eq. (8.1), which consists of two equations and two unknowns:

$$V_{1}\left(\frac{1}{2\Omega} + \frac{1}{12\Omega}\right) - V_{2}\left(\frac{1}{12\Omega}\right) = +4 \text{ A}$$

$$V_{2}\left(\frac{1}{12\Omega} + \frac{1}{6\Omega}\right) - V_{1}\left(\frac{1}{12\Omega}\right) = -2 \text{ A}$$
(8.1)

producing

$$\frac{7}{12}V_1 - \frac{1}{12}V_2 = +4$$

$$7V_1 - V_2 = 48$$

$$-\frac{1}{12}V_1 + \frac{3}{12}V_2 = -2$$

$$-1V_1 + 3V_2 = -24$$

1

and

$$V_{1} = \frac{\begin{vmatrix} 48 & -1 \\ -24 & 3 \end{vmatrix}}{\begin{vmatrix} 7 & -1 \\ -1 & 3 \end{vmatrix}} = \frac{120}{20} = + 6 \text{ V}$$
$$\begin{vmatrix} 7 & 48 \end{vmatrix}$$

$$V_2 = \frac{\left|-1 - 24\right|}{20} = \frac{-120}{20} = -6$$
 V

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Since  $V_1$  is greater than  $V_2$ , the current through  $R_3$  passes from  $V_1$  to  $V_2$ . Its value is

$$I_{R_3} = \frac{V_1 - V_2}{R_3} = \frac{6 \text{ V} - (-6 \text{ V})}{12 \Omega} = \frac{12 \text{ V}}{12 \Omega} = 1 \text{ A}$$

The fact that  $V_1$  is positive results in a current  $I_{R_1}$  from  $V_1$  to ground equal to

$$I_{R_1} = \frac{V_{R_1}}{R_1} = \frac{V_1}{R_1} = \frac{6 \text{ V}}{2 \Omega} = 3 \text{ A}$$

Finally, since  $V_2$  is negative, the current  $I_{R_2}$  flows from ground to  $V_2$  and is equal to

$$I_{R_2} = \frac{V_{R_2}}{R_2} = \frac{V_2}{R_2} = \frac{6 \text{ V}}{6 \Omega} = 1 \text{ A}$$

8.12 (Star-Delta) Y-  $\Delta$  (T-  $\pi$ ) AND  $\Delta$  -Y ( $\pi$  -T) CONVERSIONS

![](_page_19_Figure_1.jpeg)

It is our purpose to find some expression for R1, R2, and R3 in terms of RA, RB, and Rc, and vice versa, that will ensure that the resistance between any two terminals of the Y configuration will be the same with the  $\Delta$  configuration inserted in place of the Y configuration (and vice versa).

![](_page_19_Figure_3.jpeg)

![](_page_20_Figure_0.jpeg)

Finding the resistance  $R_{a-c}$  for the Y and  $\Delta$  configurations.

 $R_{a-c}\left(\mathbf{Y}\right) = R_{a-c}\left(\Delta\right)$ 

$$R_{a-c} = R_1 + R_3 = \frac{R_B(R_A + R_C)}{R_B + (R_A + R_C)}$$
(8.3a)

$$R_{a-b} = R_1 + R_2 = \frac{R_C(R_A + R_B)}{R_C + (R_A + R_B)}$$

$$R_{b-c} = R_2 + R_3 = \frac{R_A(R_B + R_C)}{R_A + (R_B + R_C)}$$

(8.3c)

(8.3b)

$$R_{1} = \frac{R_{B}R_{C}}{R_{A} + R_{B} + R_{C}} \qquad R_{2} = \frac{R_{A}R_{C}}{R_{A} + R_{B} + R_{C}} \qquad R_{3} = \frac{R_{A}R_{B}}{R_{A} + R_{B} + R_{C}}$$

Note that each resistor of the Y is equal to the product of the resistors in the two closest branches of the  $\Delta$  divided by the sum of the resistors in the  $\Delta$ .

$$R_{A} = \frac{R_{1}R_{2} + R_{1}R_{3} + R_{2}R_{3}}{R_{1}} \qquad R_{B} = \frac{R_{1}R_{2} + R_{1}R_{3} + R_{2}R_{3}}{R_{2}} \qquad R_{C} = \frac{R_{1}R_{2} + R_{1}R_{3} + R_{2}R_{3}}{R_{3}}$$

Note that the value of each resistor of the  $\Delta$  is equal to the sum of the possible product combinations of the resistances of the Y divided by the resistance of the Y farthest from the resistor to be determined.

Let us consider what would occur if all the values of a  $\Delta$  or Y were the same. If RA RB Rc, (using RA only) the following:

$$R_{3} = \frac{R_{A}R_{B}}{R_{A} + R_{B} + R_{C}} = \frac{R_{A}R_{A}}{R_{A} + R_{A} + R_{A}} = \frac{R_{A}^{2}}{3R_{A}} = \frac{R_{A}}{3}$$

and, following the same procedure,

$$R_1 = \frac{R_A}{3} \qquad R_2 = \frac{R_A}{3}$$

In general, therefore, 
$$R_{\rm Y} = \frac{R_{\Delta}}{3}$$
  $R_{\Delta} = 3R_{\rm Y}$ 

The Y and the  $\Delta$  often appear as shown below. They are then referred to as a tee (T) and a pi ( $\pi$ ) network, respectively. The equations used to convert from one form to the other are exactly the same as those developed for the Y and  $\Delta$  transformation.

![](_page_22_Figure_2.jpeg)

The relationship between the Y and T configurations and the  $\Delta$  and  $\pi$  configurations.

**EXAMPLE 8.29** Find the total resistance of the network shown, where  $R_A = 3\Omega$ ,  $R_B = 3\Omega$ , and  $R_C = 6\Omega$ .

$$R_{1} = \frac{R_{B}R_{C}}{R_{A} + R_{B} + R_{C}} = \frac{(3\ \Omega)(6\ \Omega)}{3\ \Omega + 3\ \Omega + 6\ \Omega} = \frac{18\ \Omega}{12} = 1.5\ \Omega \leftarrow R_{T}$$

$$R_{2} = \frac{R_{A}R_{C}}{R_{A} + R_{B} + R_{C}} = \frac{(3\ \Omega)(6\ \Omega)}{12\ \Omega} = \frac{18\ \Omega}{12} = 1.5\ \Omega \leftarrow R_{T}$$

$$R_{3} = \frac{R_{A}R_{B}}{R_{A} + R_{B} + R_{C}} = \frac{(3\ \Omega)(3\ \Omega)}{12\ \Omega} = \frac{9\ \Omega}{12} = 0.75\ \Omega$$

Replacing the  $\Delta$  by the Y, as shown in Fig., yields

$$R_T = 0.75 \ \Omega + \frac{(4 \ \Omega + 1.5 \ \Omega)(2 \ \Omega + 1.5 \ \Omega)}{(4 \ \Omega + 1.5 \ \Omega) + (2 \ \Omega + 1.5 \ \Omega)}$$
  
= 0.75 \ \Omega + \frac{(5.5 \ \Omega)(3.5 \ \Omega)}{5.5 \ \Omega + 3.5 \ \Omega}  
= 0.75 \ \Omega + 2.139 \ \Omega   
$$R_T = 2.89 \ \Omega$$

![](_page_23_Figure_4.jpeg)

**EXAMPLE 8.30** Find the total resistance of the network shown.

a. Converting the  $\Delta$  to a Y:

$$R_{\rm Y} = \frac{R_{\Delta}}{3} = \frac{6 \,\Omega}{3} = 2 \,\Omega$$

The network then appears as shown

$$R_T = 2 \left[ \frac{(2 \Omega)(9 \Omega)}{2 \Omega + 9 \Omega} \right] = 3.27 \Omega$$

![](_page_24_Figure_5.jpeg)

Converting the  $\Delta$  configuration of Fig. 8.91 to a Y configuration.

![](_page_24_Picture_7.jpeg)

![](_page_24_Figure_8.jpeg)

Substituting the Y configuration for the converted  $\Delta$  into the network

b. Converting the Y to a  $\Delta$  :

$$R_{\Delta} = 3R_{Y} = (3)(9 \ \Omega) = 27 \ \Omega$$

$$R'_{T} = \frac{(6 \ \Omega)(27 \ \Omega)}{6 \ \Omega + 27 \ \Omega} = \frac{162 \ \Omega}{33} = 4.91 \ \Omega$$

$$R_{T} = \frac{R'_{T}(R'_{T} + R'_{T})}{R'_{T} + (R'_{T} + R'_{T})} = \frac{R'_{T}2R'_{T}}{3R'_{T}} = \frac{2R'_{T}}{3}$$

$$= \frac{2(4.91 \ \Omega)}{3} = 3.27 \ \Omega$$

![](_page_25_Figure_2.jpeg)

# **PROBLEMS**

SECTION 8.2 Current Sources: 1, 2, 4, 6

SECTION 8.3 Source Conversions: 7, 8, 9, 10

SECTION 8.4 Current Sources in Parallel: 11, 12, 14

SECTION 8.6 Branch-Current Analysis: 15, 16, 18, 24

SECTION 8.9 Nodal Analysis: 35, 36, 38, 40

SECTION 8.12 Y –  $\Delta$  (T- $\pi$ ) and  $\Delta$  - Y ( $\pi$ -T) Conversions: 51, 52, 54, 56